<span id="page-0-0"></span>Learning to control unknown feedback linearizable systems from expert demonstrations

#### Alimzhan Sultangazin

March 7, 2022





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- Express comfort with a cost function and use optimal control?

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- We believe it is easier to collect demonstrations of "comfortable driving" — learning from demonstrations (LfD).
- Many other control tasks benefit from LfD, e.g., manufacturing [1]; healthcare [2]; robotics [3]. The growing research interest in LfD [4] reflects the need for a well-defined design methodology.

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 $\bullet$  In this work, we combine two methodologies: the LfD methodology from [5] and the data-driven control methodology from [6].

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- $\bullet$  In this work, we combine two methodologies: the LfD methodology from [5] and the data-driven control methodology from [6].
- The LfD methodology leverages the fact that the solution set of a linear system is a vector space to construct a stabilizing control law.

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	- ▶ Results in [5] rely on the assumption that we have complete knowledge of the system.
- The data-driven control methodology provides a method for stabilizing unknown feedback linearizable SISO systems with standard linear control techniques and sufficiently fast sampling rates.
	- $\triangleright$  Data-driven control allows us to relax the assumption on the knowledge of the system.

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- Issues with ML-based approaches: need many demonstrations, cannot recover from disturbances [10], few formal stability guarantees.

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- Issues with ML-based approaches: need many demonstrations, cannot recover from disturbances [10], few formal stability guarantees.
- $\bullet$  Control-theoretic approaches: the work in [11] is conceptually the closest to ours, but we do not assume the expert's policy to be linear.

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Consider a continuous-time control-affine system:

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\Sigma: \quad \dot{x} = f(x) + g(x)u, \quad y = h(x). \tag{1}
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Triple  $(x, u, y) : \mathbb{R}_0^+ \to \mathbb{R}^n \times \mathbb{R} \times \mathbb{R}$ : a solution of the system  $(1)$ .  $\bullet$  Functions x, u, and y are trajectory, control input, and output of [\(1\)](#page-13-0).

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### Definition

A controller  $\kappa:\mathbb{R}^n\to\mathbb{R}^m$  is *asymptotically stabilizing* for system  $(1)$  if the origin is uniformly asymptotically stable $^1$  for the system  $(1)$  with  $u=\kappa(x).$ 

 $\llbracket$   $^1$  that is, there exists a class KL function $^1$   $\beta$  such that  $\Vert x(t)\Vert \leq \beta(\Vert x(t_0)\Vert, t-t_0), \ \forall t\geq t_0\geq 0.$ 

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- Assume the expert produces  $n + 1$  expert solutions of system [\(1\)](#page-13-0) with length  $\tau \in \mathbb{R}$ :

 $\mathcal{D} = \{(\mathsf{x}^i, \mathsf{u}^i, \mathsf{y}^i)\}_{i=1}^{n+1}, \text{ with } (\mathsf{x}^i, \mathsf{u}^i, \mathsf{y}^i) : [0, \tau] \to \mathbb{R}^n \times \mathbb{R} \times \mathbb{R}.$ 

We collect the corresponding set of measurement samples:

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\mathcal{D}_s = \{y_s^i\}_{i=1}^{n+1}, \quad y_s^i(k) \triangleq y^i(k)\quad \text{(2)}
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 (2)

• Assume the system  $(1)$  is feedback linearizable with h having a relative degree *n*, i.e., for all  $x \in \mathbb{R}^n$ :

$$
L_g L_f^i h(x) = 0, \quad i = 0, \dots, n-2,
$$
 (3)

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L_g L_f^{n-1} h(x) \neq 0. \tag{4}
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In addition, assume w.l.o.g. that  $L_g L_f^{n-1}$  $_{f}^{n-1}h > 0.$ 

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- For clarity of exposition, in what follows we assume  $n = 2$ .

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- For clarity of exposition, in what follows we assume  $n = 2$ .
- We can rewrite the nonlinear system dynamics [\(1\)](#page-13-0) in the coordinates:

$$
z = \Phi(x) = \begin{bmatrix} h(x) & L_f h(x) \end{bmatrix}^T, \tag{5}
$$

and get:

$$
\begin{aligned}\n\dot{z}_1 &= z_2, \\
\dot{z}_2 &= \alpha(z) + \beta(z)u = w, \\
y &= z_1\n\end{aligned} \tag{6}
$$

where  $\alpha=(L_f^2 h)\circ \Phi^{-1}$ ,  $\beta=(L_{\bf g}L_f h)\circ \Phi^{-1}$ , and  $w\triangleq \alpha(z)+\beta(z)u$ .

# Learning from demonstrations when dynamics are known

• For now, assume that we know functions  $\alpha$  and  $\beta$  and are given the set  $\mathcal{D}_{\mathsf{e}} = \{(\mathsf{z}^i,w^i)\}$ , where  $\mathsf{z}^i = \Phi(\mathsf{x}^i)$  and  $w^i = \alpha(\mathsf{z}^i) + \beta(\mathsf{z}^i)u^i$ .

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$$
Z(t) \triangleq [z2(t) - z1(t) | \cdots | zn+1(t) - z1(t)] \qquad (7)
$$
  
 
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W(t) \triangleq [w2(t) - w1(t) | \cdots | wn+1(t) - w1(t)]. \qquad (8)
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Lemma (Affine comb. of inputs  $\Rightarrow$  Affine comb. of trajectories)

- Suppose we are given a set of finite-length solutions  $\{(z^i, w^i)\}_{i=1}^{n+1}$  of the system [\(6\)](#page-22-0), where each  $(z^i, w^i)$  is defined for  $0 \le t \le \tau$ ,  $\tau \in \mathbb{R}$ .
- Assume that  $\{z^i(0)\}_{i=1}^{n+1}$  is an affinely independent set.

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Assume that  $\{z^i(0)\}_{i=1}^{n+1}$  is an affinely independent set.

Then, for  $t_0 \le t \le \tau + t_0$ , under the input  $w(t) = W(t - t_0) \zeta$  with  $\zeta = Z^{-1}(0)z_0$ , the solution of the system  $(6)$  with  $z(t_0) = z_0$  is:

$$
z(t)=Z(t-t_0)\zeta.
$$







$$
\begin{cases} 1. & \text{For } t \in [0, \tau), \\ w(t) = W(t)\zeta(0) \text{ with } \\ \zeta(0) = Z^{-1}(0)z(0) \end{cases}
$$

2. For 
$$
t \in [\tau, 2\tau)
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3. For 
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t \in [2\tau, 3\tau)
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\n $w(t) = W(t - 2\tau)\zeta(2)$  with  
\n $\zeta(2) = Z^{-1}(0)z(2\tau)$ 

We apply the following preliminary controller:

$$
u(t) = \beta^{-1}(z(t))(-\alpha(z(t)) + w(t, z(t))), \qquad (9)
$$

to bring the system  $(1)$  to the form  $y^{(n)}=w$  and use the control law:

<span id="page-34-0"></span>
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w(t, z(t)) = W(t - p\tau)Z^{-1}(t - p\tau)z(t) = K(t)z(t),
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for all  $t \in [p\tau, (p+1)\tau)$  and  $p \in \mathbb{N}_0$ .

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### Lemma (Minimal length of demonstrations for stability)

Suppose a set  $\mathcal{D}_e = \{(z^i,w^i)\}_{i=1}^{n+1}$  of solutions with length  $\tau \in \mathbb{R}^+$  is generated by the system [\(6\)](#page-22-0) in closed loop with an asymptotically stabilizing controller  $u = \kappa(z)$ .

Assume that  $\{z^i(t)\}_{i=1}^{n+1}$  is affinely independent for all  $t\in [0,\tau].$ Then, there is  $\bar{\tau} \in \mathbb{R}^+$  such that for all  $\tau \geq \bar{\tau}$ , the origin of the system [\(6\)](#page-22-0) in closed loop with the controller in [\(10\)](#page-34-0) is uniformly exponentially stable.

# Learning from demonstrations when dynamics are unknown (using data-driven control)

Previous assumptions: know  $\mathcal{D}_e = \{(z^i, w^i)\}_{i=1}^{n+1}$  and functions  $\alpha$  and  $\beta$ .

[6] Fraile, Marchi, and Tabuada, 2020, "Data-driven Stabilization of SISO Feedback Linearizable Systems"

Previous assumptions: know  $\mathcal{D}_e = \{(z^i, w^i)\}_{i=1}^{n+1}$  and functions  $\alpha$  and  $\beta$ . The data-driven control from [6] consists of two components: state estimator and dynamic controller.

[6] Fraile, Marchi, and Tabuada, 2020, "Data-driven Stabilization of SISO Feedback Linearizable Systems"

- Previous assumptions: know  $\mathcal{D}_e = \{(z^i, w^i)\}_{i=1}^{n+1}$  and functions  $\alpha$  and  $\beta$ .
- The data-driven control from [6] consists of two components: state estimator and dynamic controller.
- We use the state estimator from [6] to:
	- ▶ estimate the set  $\widehat{\mathcal{D}}_e = \{(\widehat{z}_s^i, \widehat{w}_s^i)\}_{i=1}^{n+1}$  from the given data  $\mathcal{D}_s = \{y_s^i\}_{i=1}^{n+1}$ ;

<sup>[6]</sup> Fraile, Marchi, and Tabuada, 2020, "Data-driven Stabilization of SISO Feedback Linearizable Systems"

Previous assumptions: know  $\mathcal{D}_e = \{(z^i, w^i)\}_{i=1}^{n+1}$  and functions  $\alpha$  and  $\beta$ .

- The data-driven control from [6] consists of two components: state estimator and dynamic controller.
- We use the state estimator from [6] to:
	- Experiment the set  $\widehat{\mathcal{D}}_e = \{(\widehat{z}_s^i, \widehat{w}_s^i)\}_{i=1}^{n+1}$  from the given data  $\mathcal{D}_s = \{y_s^i\}_{i=1}^{n+1}$ ;
	- **•** provide estimates of  $\hat{z}_s$  and  $\hat{w}_s$  to the dynamic controller.

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	- **•** provide estimates of  $\hat{z}_s$  and  $\hat{w}_s$  to the dynamic controller.
- The dynamic controller from [6] tracks the virtual input w from the learned controller [\(10\)](#page-34-0) without knowing  $\alpha$  and  $\beta$ .

[6] Fraile, Marchi, and Tabuada, 2020, "Data-driven Stabilization of SISO Feedback Linearizable Systems"

# The main result

#### Theorem

Consider an unknown SISO system [\(1\)](#page-13-0) where h has a relative degree n.

- Let T be the sampling time and  $\tau$  be the demonstration length.
- Suppose we are given  $\mathcal{D}_s = \{y_s^i\}_{i=1}^{n+1}$  generated by the system  $(1)$  in closed loop with a stabilizing expert. and the state estimator from [6] is used to construct  $\widehat{\mathcal{D}}_e = \left\{ \left( \widehat{z}^i_s, \widehat{w}^i_s \right) \right\}_{i=1}^{n+1}$ . Then, there exist  $\bar{T} \in \mathbb{R}^+$  and  $\bar{\tau} \in \mathbb{R}^+$  so that, for any  $T \in [0, \bar{T}]$  and any  $\tau > \bar{\tau}$ , the dynamic controller, based on the learned controller [\(10\)](#page-34-0), renders the closed-loop solutions bounded and  $\lim_{t\to\infty} x(t) = 0$ .

# Conclusion

We have extended a methodology in [5] for constructing a stabilizing controller from expert demonstrations to unknown SISO systems.

[5] Sultangazin et al., 2021, Watch and Learn: Learning to control feedback linearizable systems from expert demonstrations

# Conclusion

- We have extended a methodology in [5] for constructing a stabilizing controller from expert demonstrations to unknown SISO systems.
- Compared to machine-learning approaches, this methodology requires few demonstrations (i.e., the minimal number of demonstrations is  $n + 1$ ) and provides formal stability guarantees.

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# Conclusion

- We have extended a methodology in [5] for constructing a stabilizing controller from expert demonstrations to unknown SISO systems.
- Compared to machine-learning approaches, this methodology requires few demonstrations (i.e., the minimal number of demonstrations is  $n + 1$ ) and provides formal stability guarantees.
- As part of future work, we plan to:
	- ▶ apply a similar methodology to learn control of unknown MIMO systems;
	- $\triangleright$  experimentally verify this methodology using the testbed with Crazyflie quadrotors we have in our laboratory.

[5] Sultangazin et al., 2021, Watch and Learn: Learning to control feedback linearizable systems from expert demonstrations